

Stable adiabatic circuit using advanced series capacitors and time variation of energy dissipation

Shunji Nakata^{1a)}, Shin'ichiro Mutoh¹, Hiroshi Makino²,
Masayuki Miyama³, and Yoshio Matsuda³

¹ NTT Microsystem Integration Laboratories

3-1 Morinosato Wakamiya, Atsugi, Kanagawa 243-0198, Japan

² Faculty of Information Science and Technology, Osaka Institute of Technology
Hirakata, Osaka 573-0196, Japan

³ Graduate School of Natural Science, Kanazawa University
Kanazawa, Ishikawa 920-1192, Japan

a) nakata@aecl.ntt.co.jp

Abstract: We discuss the stability of an adiabatic stepwise-charging circuit with advanced series capacitors, which is effective for the reduction of the applied voltage to each capacitor. SPICE simulation shows that this circuit is stable even if the initial voltages are lower than zero. For the analytical discussion, we derive a matrix that connects charge and voltage in the circuit and show that the matrix is a positive-definite symmetric one. Therefore, the step voltage is generated spontaneously. We also derive energy dissipation analytically using tank capacitor voltage. Using this formula and SPICE simulation, we clarify that energy dissipation decreases monotonically as a function of time and finally reaches the minimum value.

Keywords: adiabatic charging, charge recycling, series capacitors, a positive-definite symmetric matrix, stability, energy dissipation

Classification: Integrated circuits

References

- [1] L. J. Svensson and J. G. Koller, "Driving a capacitive load without dissipating fCV^2 ," *Proc. IEEE Symp. Low Power Electron.*, pp. 100–101, 1994.
- [2] R. Lal, W. Athas, and L. Svensson, "A low power adiabatic driver system for AMLCDs," *Proc. IEEE Symp. VLSI Circuits*, pp. 198–201, 2000.
- [3] S. Nakata, "Stability of adiabatic circuit using asymmetric 1D-capacitor array between the power supply and ground," *IEICE Electron. Express*, vol. 4, no. 5, pp. 165–171, 2007.
- [4] M. van Elzaker et al., "A 1.9 μ W 4.4 fJ/conversion-step 10b 1 MS/s Charge-Redistribution ADC," *Proc. ISSCC Dig.*, pp. 244–245, 2008.
- [5] S. Nakata, T. Kusumoto, M. Miyama, and Y. Matsuda, "Adiabatic SRAM

- with a Large Margin of V_T Variation by Controlling the Cell-Power-Line and Word-Line Voltage,” *Proc. IEEE ISCAS*, pp. 393–396, 2009.
- [6] S. Nakata, Y. Katagiri, and S. Matsuno, “Electrostatic energy, potential energy and energy dissipation for a width-variable capacitor system during adiabatic charging,” *J. Appl. Phys.*, vol. 101, p. 034911, 2007.

1 Introduction

Reducing the power dissipation of circuits is an important issue. A charge recycling regenerator with a switched capacitor circuit is one of the most promising solutions and has been researched for adiabatic logic [1, 2, 3, 4]. In a previous article [5], we proposed a switched capacitor circuit with advanced series capacitors and showed by SPICE that the tank capacitor voltage converges to the step voltage spontaneously when the initial voltages are larger than or equal to zero. However, it is not clear whether the circuit is stable when they are negative (i.e., below the value of GND), which is often caused by external noise.

In this article, we confirm by SPICE that the circuit is stable even if the initial voltages are negative and that the circuit reaches the stable state five times as rapidly as the conventional one [1]. The stability of this circuit is proved generally by an analytical method. We also discuss the energy dissipation as a function of time. It is clarified that, although the voltages of the tank capacitor change variously (sometimes increase and sometimes decrease), the energy dissipation always decreases monotonically and finally reaches the minimum value.

2 Stability of the regenerator with series capacitors

The conventional switched capacitor regenerator circuit and the regenerator with the advanced series capacitors are shown in Figs. 1 (a) and (b). This series capacitors circuit is advanced compared to the previous series one [3] because, in the four-step case, the number of tank capacitors C_i decreases from 4 to 3, which is the same as in Fig. 1 (a). Therefore, we can reduce

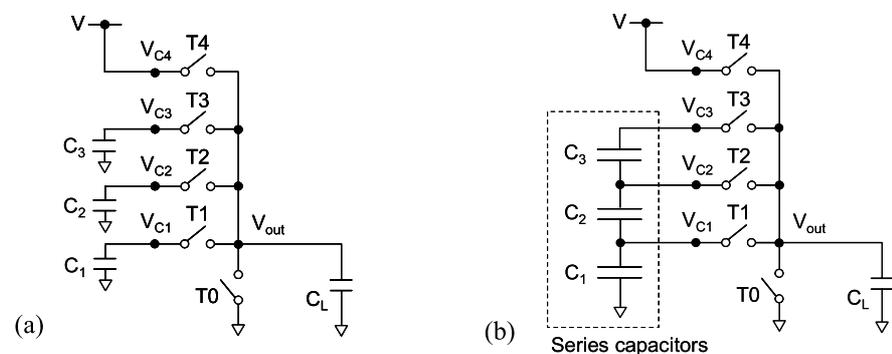


Fig. 1. Switched capacitor circuit. (a) Conventional circuit. (b) Advanced series capacitors circuit.

the number of capacitors by using the advanced one. In Figs. 1 (a) and (b), C_L is load capacitance, V is the power supply voltage, V_{out} is output voltage, and V_{C_i} is the voltage of the node connected to the upper plate of C_i . The switching transistor is a parallel connection of pMOSFETs and nMOSFETs. T0, T1, T2, T3, T4, T3, T2, and T1 turn on successively and this operation is repeated.

The circuit simulation results are shown in Fig. 2. We used the 0.25- μm design rule. Threshold voltages were 0.4 and -0.4 V in the nMOS and pMOS transistors, respectively. C_1 , C_2 , and C_3 were the same value: 100 pF. C_L was 0.4 pF. The period of the four-step waveform cycle was 0.2 μs . The initial V_{C_1} , V_{C_2} , and V_{C_3} values were set to -0.4 V. The gate width was 6 μm . The gate length and V are 0.25 μm and 2 V in Fig. 2 (a), and 0.5 μm and 4 V in Fig. 2 (b), respectively. In Fig. 2, the blue and red lines show the advanced series circuit and the conventional one, respectively. In both cases, after 200 μs , V_{C_i} becomes $iV/4$ spontaneously. From the results, it is clear that the advanced series circuit is very stable even if the initial V_{C_i} is negative due to external noise and that it reaches the stable state five times as rapidly as the conventional one. Another feature of the advanced series circuit is that the voltage of the capacitors is smaller than $V/4$ due to the series connection. On the other hand, for the conventional one, the maximum voltage of the capacitors is $3V/4$. This difference is a serious problem when we use an electric double layer capacitor (EDLC). The endurance voltage of the EDLC is 2.5 V so that we cannot use the conventional circuit with $V = 4$ V because V_{C_3} in Fig. 1 (a) reaches 3 V as shown in Fig. 2 (b).

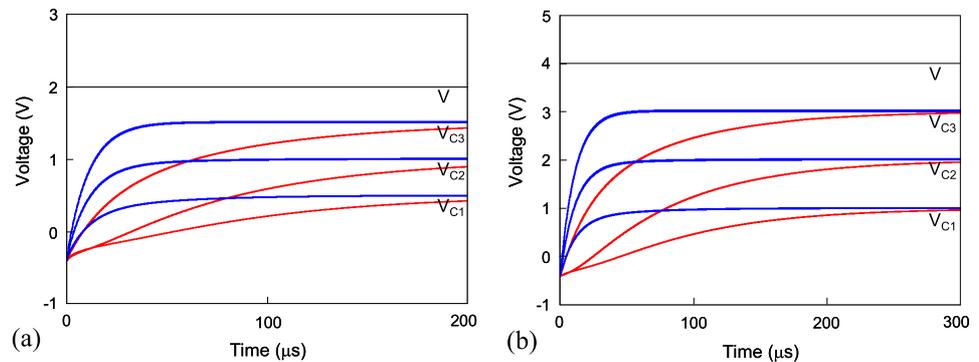


Fig. 2. Voltage change of V_{C_i} when V is (a) 2 and (b) 4 V. The blue and red lines are for the advanced series circuit and conventional one, respectively.

Next, we investigate the reason for the stability generally by using an analytical method. Here, we assume that $C_i \gg C_L$. The y_i is the node connected to the upper plate of C_i . Let Q_{ti} be the transferred charge quantity from y_i to C_L at the i th step voltage [Fig. 3 (a)] and Q_{ri} be the restored charge from C_L to y_i [Fig. 3 (b)]. We define Q_i as the amount of charge stored in the capacitor plates connected to y_i . Then, assuming that the number of the steps is N , $V_{C_0} = 0$, and $V_{C_N} = V$, ΔQ_i (the change of Q_i) after charging

and restoring can be written as [3]

$$\Delta Q_i = -Q_{ti} + Q_{ri} = C_L(V_{C(i-1)} - 2V_{C_i} + V_{C(i+1)}), \quad (1 \leq i \leq N-1). \quad (1)$$

Here, we define V_i as $V_i = V_{C_i} - iV/N$. Using V_i and (1), we have

$$\Delta Q_i = C_L(V_{i-1} - 2V_i + V_{i+1}), \quad (1 \leq i \leq N-1). \quad (2)$$

Next, we define v_i as the voltage difference between the capacitor plates [Fig. 3 (a)]. Then, using $V_{C_i} = v_1 + v_2 + \dots + v_{i-1} + v_i$, we have

$$\begin{bmatrix} V_{C_1} \\ \vdots \\ V_{C(N-1)} \end{bmatrix} = \mathbf{B} \begin{bmatrix} v_1 \\ \vdots \\ v_{N-1} \end{bmatrix}, \quad \text{where } \mathbf{B} = \begin{bmatrix} 1 & & 0 \\ & \ddots & \\ & & 1 \end{bmatrix}. \quad (3)$$

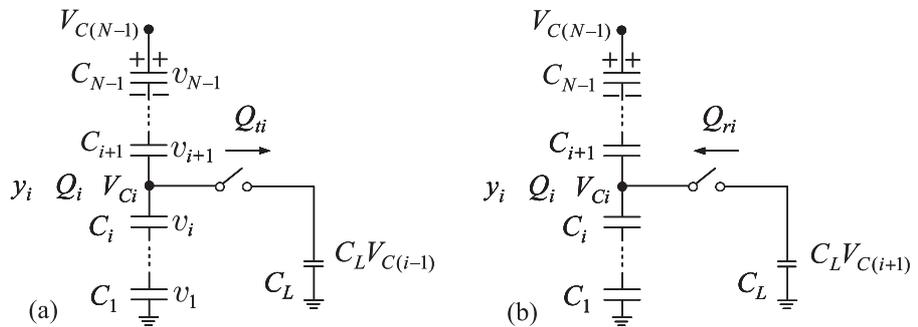


Fig. 3. Definitions of charge and voltage in the regenerator. (a) Q_{ti} is transferred from y_i to C_L at the i th step. (b) Q_{ri} is restored from C_L to y_i at the i th step.

From Fig. 3, we have

$$\begin{bmatrix} Q_1 \\ Q_2 \\ \vdots \\ Q_{N-1} \end{bmatrix} = \mathbf{D} \begin{bmatrix} v_1 \\ v_2 \\ \vdots \\ v_{N-1} \end{bmatrix}, \quad \text{where } \mathbf{D} = \begin{bmatrix} C_1 & -C_2 & & 0 \\ & C_2 & -C_3 & \\ & & \ddots & \ddots \\ & & & C_{N-2} & -C_{N-1} \\ 0 & & & & C_{N-1} \end{bmatrix}. \quad (4)$$

Then, using (3) and (4), and considering the difference in Q_i and V_{C_i} after one cycle operation, we have

$$\begin{bmatrix} \Delta Q_1 \\ \vdots \\ \Delta Q_{N-1} \end{bmatrix} = \mathbf{D}\mathbf{B}^{-1} \begin{bmatrix} \Delta V_{C_1} \\ \vdots \\ \Delta V_{C(N-1)} \end{bmatrix}, \quad (5)$$

where ΔV_{C_i} is the change of V_{C_i} after charging and restoring. Using (5) and $\Delta V_{C_i} = \Delta V_i$, we have

$$\begin{bmatrix} \Delta Q_1 \\ \vdots \\ \Delta Q_{N-1} \end{bmatrix} = \mathbf{F} \cdot \begin{bmatrix} \Delta V_1 \\ \vdots \\ \Delta V_{N-1} \end{bmatrix}, \quad \text{where } \mathbf{F} = \mathbf{D}\mathbf{B}^{-1}. \quad (6)$$

The B^{-1} is calculated as in ref. 3. Therefore, we have

$$F = DB^{-1} = \begin{bmatrix} C_1 + C_2 & -C_2 & & & & \\ -C_2 & C_2 + C_3 & -C_3 & & & \\ & -C_3 & \ddots & \ddots & & \\ & & \ddots & C_{N-2} + C_{N-1} & -C_{N-1} & \\ & & & -C_{N-1} & C_{N-1} & \end{bmatrix}. \quad (7)$$

Using (7), we have

$$\mathbf{x}^t \mathbf{F} \mathbf{x} = C_1 x_1^2 + C_2 (x_1 - x_2)^2 + \cdots + C_{N-1} (x_{N-2} - x_{N-1})^2, \quad (8)$$

where \mathbf{x} is one of any vector. While the \mathbf{F} is different from that in ref. 3, we easily find that it is a positive-definite symmetric matrix so that step voltage is generated spontaneously using the theory in ref. 3.

3 Time variation of energy dissipation

Next, we investigate how the energy dissipation changes as a function of time. In the four-step case, the work done by the regenerator during charging W_1 is written as [6]

$$\begin{aligned} W_1 &= V_{C1} \Delta Q_1 + V_{C2} \Delta Q_2 + V_{C3} \Delta Q_3 + V_{C4} \Delta Q_4 \\ &= C_L [V_{C1} V_{C1} + V_{C2} (V_{C2} - V_{C1}) + V_{C3} (V_{C3} - V_{C2}) + V_{C4} (V_{C4} - V_{C3})]. \end{aligned} \quad (9)$$

The work done by the regenerator when restoring W_2 is written as

$$\begin{aligned} W_2 &= -V_{C1} \Delta Q_2 - V_{C2} \Delta Q_3 - V_{C3} \Delta Q_4 \\ &= -C_L [V_{C1} (V_{C2} - V_{C1}) + V_{C2} (V_{C3} - V_{C2}) + V_{C3} (V_{C4} - V_{C3})]. \end{aligned} \quad (10)$$

W_2 is negative, which means the regenerator gets energy from C_L . Using the energy conservation law, we have

$$W_1 = E_{diss1} + U, \quad (11)$$

where E_{diss1} is the energy dissipation during charging and U is the electrostatic energy of a load capacitor. We also have

$$U = -W_2 + E_{diss2}, \quad (12)$$

where E_{diss2} is the energy dissipation during restoring. Therefore, using (11) and (12), we have

$$W_1 + W_2 = E_{diss1} + E_{diss2} = E_{diss}, \quad (13)$$

where E_{diss} is the total energy dissipation during one cycle. Then, using (9) and (10), we have

$$E_{diss} = C_L [V_{C1}^2 + (V_{C2} - V_{C1})^2 + (V_{C3} - V_{C2})^2 + (V_{C4} - V_{C3})^2]. \quad (14)$$

The minimum of E_{diss} is calculated using the method of Lagrange multipliers. We denote q_1 , q_2 , q_3 , and q_4 as

$$q_1 = V_{C1}, \quad q_2 = V_{C2} - V_{C1}, \quad q_3 = V_{C3} - V_{C2}, \quad q_4 = V_{C4} - V_{C3}. \quad (15)$$

Then, we have $q_1 + q_2 + q_3 + q_4 = V$. We define L as

$$L = E_{diss} - \lambda(q_1 + q_2 + q_3 + q_4 - V). \quad (16)$$

By calculating $\partial L/\partial q_i = 0$ and $\partial L/\partial \lambda = 0$, we have $q_1 = q_2 = q_3 = q_4$ easily, which means $V_{C_i} = iV/4$. Therefore, it is clarified that E_{diss} takes the minimum when the step voltage is generated. Regarding the time variation of E_{diss} , we can calculate this value using SPICE.

The simulation result is shown in Fig. 4(a). The lower lines show E_{diss} of the advanced series circuit (blue) and the conventional one (red) with the condition in Fig. 2(a). The upper ones show those with almost the same condition but with $V_{C1} = 2.4$, $V_{C2} = -0.4$, $V_{C3} = 1.0$ V, and C_1 , C_2 , and C_3 values of 100, 50, and 100 pF, respectively. First, we discuss the lower lines. When $t = 0$ s, E_{diss}/C_L is equal to 5.9, which is valid from the initial condition. When $t = 200 \mu\text{s}$, E_{diss}/C_L is equal to 1, which is also valid from the final state such that $V_{C_i} = iV/4$. The E_{diss} in the advanced series circuit decreases more rapidly than in the conventional one. This means that the circuit reaches the stable state more rapidly than the conventional one.

Next, we consider the upper lines. The change of V_{C_i} with this condition is shown in Fig. 4(b). The blue and red lines in Fig. 4(b) are V_{C_i} of the advanced series circuit and the conventional one, respectively. In the upper lines in Fig. 4(a), E_{diss} in the conventional circuit (red) decreases more rapidly than in the advanced series one (blue) at this time. However, this initial V_{C_i} is a very rare case and would hardly ever occur. Normally, the initial V_{C_i} is around zero as in Fig. 2(a), at which we can easily confirm that E_{diss} in the

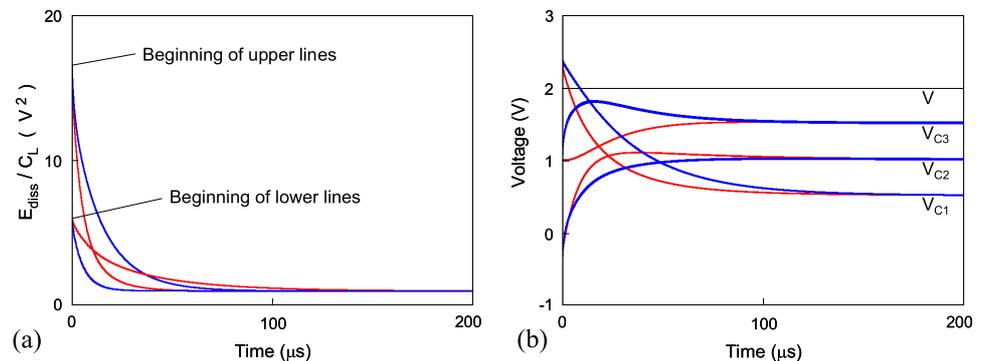


Fig. 4. (a) Time variation of energy dissipation during one cycle in the circuit with advanced series capacitors (blue) and with the conventional ones (red). (b) Change of V_{C_i} in the advanced series circuit (blue) and the conventional one (red) when the initial V_{C1} , V_{C2} , and V_{C3} are 2.4, -0.4 , and 1.0 V, respectively.

advanced series circuit decreases more rapidly than that in the conventional one. This means the stable state in the advanced series circuit is generated more rapidly. We can say that this is the merit of the advanced series circuit.

Regarding the time variation of E_{diss} , interestingly, E_{diss} in both circuits decreases monotonically and reaches the minimum value, even if the voltages change variously as in Fig. 4 (b). This means, in other words, that the voltage state in the circuit proceeds in a direction such that the energy dissipation becomes smaller and finally reaches the minimum value. This phenomenon in the circuit reminds us of the other principle in the electromagnetic theory; namely, that the steady-state current distribution in a conductor always satisfies the condition such that the energy dissipation (Joule energy) is the minimum value.

4 Conclusion

In summary, we analyzed an adiabatic circuit with advanced series capacitors. We confirmed by SPICE that this circuit is stable even if the initial voltages are negative and proved its stability generally by an analytical method. We also clarified that the voltage state in the circuit proceeds in a direction such that the energy dissipation becomes smaller and finally reaches the minimum value.